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# Model Reduction for Dynamical Systems

— Lecture 4 —

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#### Introduction

- Model Reduction for Dynamical Systems
- Application Areas
- Motivating Examples



#### Mathematical Basics

- Numerical Linear Algebra
- Systems and Control Theory
- Qualitative and Quantitative Study of the Approximation Error



#### Model Reduction by Projection

- Introduction
- Projection and Interpolation

#### Modal Truncation

- Basic Principle
- Dominant Pole Algorithm

Outline		

#### Introduction

#### Mathematical Basics

- 3 Model Reduction by Projection
  - Introduction
  - Projection and Interpolation



- Automatic generation of compact models.
- Satisfy desired error tolerance for all admissible input signals, i.e., want

 $||y - \hat{y}|| < \text{tolerance} \cdot ||u|| \qquad \forall u \in L_2(\mathbb{R}, \mathbb{R}^m).$ 

 $\implies$  Need computable error bound/estimate!

- Preserve physical properties:
  - stability (poles of G in  $\mathbb{C}^-$ ),
  - minimum phase (zeroes of G in  $\mathbb{C}^-$ ),
  - passivity

 $\int_{-\infty}^{t} u(\tau)^{\mathsf{T}} y(\tau) \, d\tau \ge 0 \quad \forall t \in \mathbb{R}, \quad \forall u \in L_2(\mathbb{R}, \mathbb{R}^m).$ 

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Introduction

Mathematical Basics

MOR by Projection

Modal Truncation

#### Model Reduction by Projection Projection Basics

#### Definition 3.1 (Projector)

A projector is a matrix  $P \in \mathbb{R}^{n \times n}$  with  $P^2 = P$ . Let  $\mathcal{V} = \operatorname{range}(P)$ , then P is projector onto  $\mathcal{V}$ . On the other hand, if  $\{v_1, \ldots, v_r\}$  is a basis of  $\mathcal{V}$  and  $V = [v_1, \ldots, v_r]$ , then  $P = V(V^T V)^{-1} V^T$  is a projector onto  $\mathcal{V}$ .

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#### Lemma 3.2 (Projector Properties)

- If P = P<sup>T</sup>, then P is an orthogonal projector (aka: Galerkin projection), otherwise an oblique projector (aka: Petrov-Galerkin projection).
- *P* is the identity operator on  $\mathcal{V}$ , i.e.,  $Pv = v \ \forall v \in \mathcal{V}$ .
- I P is the complementary projector onto ker P.
- If  $\mathcal{V}$  is an A-invariant subspace corresponding to a subset of A's spectrum, then we call P a spectral projector.

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#### Methods:

- Modal Truncation
- Balanced Truncation
- Rational Interpolation (Padé-Approximation and (rational) Krylov Subspace Methods)
- many more...

Joint feature of these methods:

computation of reduced-order model (ROM) by projection!

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range 
$$(V) = \mathcal{V}$$
, range  $(W) = \mathcal{W}$ ,  $W^T V = I_r$ .

Then, with  $\hat{x} = W^T x$ , we obtain  $x \approx V \hat{x}$  so that

$$\|x-\tilde{x}\|=\|x-V\hat{x}\|,$$

and the reduced-order model is

$$\hat{A} := W^T A V, \quad \hat{B} := W^T B, \quad \hat{C} := C V, \quad (\hat{D} := D).$$

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$$\hat{A} := W^{\mathsf{T}} A V, \quad \hat{B} := W^{\mathsf{T}} B, \quad \hat{C} := C V, \quad (\hat{D} := D).$$

Important observation:

• The state equation residual satisfies  $\dot{\tilde{x}} - A\tilde{x} - Bu \perp W$ , since

$$W^{T}\left(\dot{\tilde{x}} - A\tilde{x} - Bu\right) = W^{T}\left(VW^{T}\dot{x} - AVW^{T}x - Bu\right)$$

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$$= \dot{\hat{x}} - \hat{A}\hat{x} - \hat{B}u = 0.$$

#### Projection ~> Rational Interpolation

Given the ROM

$$\hat{A} = W^T A V, \quad \hat{B} = W^T B, \quad \hat{C} = C V, \quad (\hat{D} = D),$$

the error transfer function can be written as

$$G(s) - \hat{G}(s) = (C(sI_n - A)^{-1}B + D) - (\hat{C}(sI_r - \hat{A})^{-1}\hat{B} + \hat{D})$$

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=  $C\left(I_n - \underbrace{V(sI_r - \hat{A})^{-1}W^T(sI_n - A)}_{=:P(s)}\right)(sI_n - A)^{-1}B.$ 

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If  $s_* \in \mathbb{C} \setminus (\Lambda(A) \cup \Lambda(\hat{A}))$ , then  $P(s_*)$  is a projector onto  $\mathcal{V}$ : range  $(P(s_*)) \subset$  range (V), all matrices have full rank  $\Rightarrow$  "=",  $P(s_*)^2 = V(s_*l_r - \hat{A})^{-1}W^T(s_*l_n - A)V(s_*l_r - \hat{A})^{-1}W^T(s_*l_n - A)$ 

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if 
$$(s_*I_n - A)^{-1}B \in \mathcal{V}$$
, then  $(I_n - P(s_*))(s_*I_n - A)^{-1}B = 0$ ,

hence

$$G(s_*) - \hat{G}(s_*) = 0 \ \Rightarrow \ G(s_*) = \hat{G}(s_*), \text{ i.e., } \hat{G} \text{ interpolates } G \text{ in } s_*!$$

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Analogously, =  $C(sl_n - A)^{-1}(l_n - (sl_n - A)V(sl_r - \hat{A})^{-1}W^T)B$ 

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If 
$$s_* \in \mathbb{C} \setminus (\Lambda(A) \cup \Lambda(\hat{A}))$$
, then  $Q(s)^H$  is a projector onto  $\mathcal{W} \Longrightarrow$   
if  $(s_*I_n - A)^{-*}C^T \in \mathcal{W}$ , then  $C(s_*I_n - A)^{-1}(I_n - Q(s_*)) = 0$ ,

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Theorem

[GRIMME '97, VILLEMAGNE/SKELTON '87]

Given the ROM

$$\hat{A} = W^T A V, \quad \hat{B} = W^T B, \quad \hat{C} = C V, \quad (\hat{D} = D),$$

and  $s_* \in \mathbb{C} \setminus (\Lambda(A) \cup \Lambda(\hat{A}))$ , if either

•  $(s_* I_n - A)^{-1} B \in range(V)$ , or

• 
$$(s_*I_n - A)^{-*}C^T \in \operatorname{range}(W),$$

then the interpolation condition

$$G(s_*)=\hat{G}(s_*).$$

in s\* holds.

Note: extension to Hermite interpolation conditions later!

		Modal Truncation ●00000
Outline		

#### Introduction

2 Mathematical Basics

3 Model Reduction by Projection



Modal Truncation

- Basic Principle
- Dominant Pole Algorithm

#### Basic method:

Assume A is diagonalizable,  $T^{-1}AT = D_A$ , project state-space onto A-invariant subspace  $\mathcal{V} = \operatorname{span}(t_1, \ldots, t_r)$ ,  $t_k = \operatorname{eigenvectors}$  corresp. to "dominant" modes / eigenvalues of A. Then with

 $V = T(:, 1:r) = [t_1, ..., t_r], \quad \tilde{W}^H = T^{-1}(1:r, :), \quad W = \tilde{W}(V^H \tilde{W})^{-1},$ 

reduced-order model is

 $\hat{A} := W^H A V = \operatorname{diag} \{\lambda_1, \dots, \lambda_r\}, \quad \hat{B} := W^H B, \quad \hat{C} = C V$ 

Also computable by truncation:

$$T^{-1}AT = \begin{bmatrix} \hat{A} \\ A_2 \end{bmatrix}, \quad T^{-1}B = \begin{bmatrix} \hat{B} \\ B_2 \end{bmatrix}, \quad CT = [\hat{C}, C_2], \quad \hat{D} = D.$$

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#### Properties:

Simple computation for large-scale systems, using, e.g., Krylov subspace methods (Lanczos, Arnoldi), Jacobi-Davidson method.

Basic method:

$$T^{-1}AT = \begin{bmatrix} \hat{A} \\ A_2 \end{bmatrix}, \quad T^{-1}B = \begin{bmatrix} \hat{B} \\ B_2 \end{bmatrix}, \quad CT = [\hat{C}, C_2], \quad \hat{D} = D.$$

## Properties:

Error bound:

$$\|G - \hat{G}\|_{\infty} \leq \|C_2\| \|B_2\| \frac{1}{\min_{\lambda \in \Lambda(A_2)} |\operatorname{Re}(\lambda)|}$$

Proof:

$$G(s) = C(sI - A)^{-1}B + D = CTT^{-1}(sI - A)^{-1}TT^{-1}B + D$$
  
=  $CT(sI - T^{-1}AT)^{-1}T^{-1}B + D$   
=  $[\hat{C}, C_2] \begin{bmatrix} (sI_r - \hat{A})^{-1} \\ (sI_{n-r} - A_2)^{-1} \end{bmatrix} \begin{bmatrix} \hat{B} \\ B_2 \end{bmatrix} + D$   
=  $\hat{G}(s) + C_2(sI_{n-r} - A_2)^{-1}B_2,$ 

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$$\|G - \hat{G}\|_{\infty} \leq \|C_2\| \|B_2\| \frac{1}{\min_{\lambda \in \Lambda(A_2)} |\operatorname{Re}(\lambda)|}$$

Proof:

$$G(s) = \hat{G}(s) + C_2(sI_{n-r} - A_2)^{-1}B_2,$$

observing that  $\|G - \hat{G}\|_{\infty} = \sup_{\omega \in \mathbb{R}} \sigma_{\max}(C_2(\jmath \omega I_{n-r} - A_2)^{-1}B_2)$ , and

$$C_2(\jmath\omega I_{n-r}-A_2)^{-1}B_2=C_2 {
m diag}\left(rac{1}{\jmath\omega-\lambda_{r+1}},\ldots,rac{1}{\jmath\omega-\lambda_n}
ight)B_2.$$

#### Basic method:

Assume A is diagonalizable,  $T^{-1}AT = D_A$ , project state-space onto A-invariant subspace  $\mathcal{V} = \operatorname{span}(t_1, \ldots, t_r)$ ,  $t_k = \operatorname{eigenvectors}$  corresp. to "dominant" modes / eigenvalues of A. Then reduced-order model is

 $\hat{A} := W^H A V = \operatorname{diag} \{\lambda_1, \dots, \lambda_r\}, \quad \hat{B} := W^H B, \quad \hat{C} = C V$ 

Also computable by truncation:

$$T^{-1}AT = \begin{bmatrix} \hat{A} \\ A_2 \end{bmatrix}, \quad T^{-1}B = \begin{bmatrix} \hat{B} \\ B_2 \end{bmatrix}, \quad CT = [\hat{C}, C_2], \quad \hat{D} = D.$$

#### Difficulties:

- Eigenvalues contain only limited system information.
- Dominance measures are difficult to compute. ([LITZ '79] use Jordan canoncial form; otherwise merely heuristic criteria, e.g., [VARGA '95]. Recent improvement: dominant pole algorithm.)
- Error bound not computable for really large-scale problems.

# Basic Principle

**BEAM**, SISO system from SLICOT Benchmark Collection for Model Reduction, n = 348, m = q = 1, reduced using 13 dominant complex conjugate eigenpairs, error bound yields  $\|G - \hat{G}\|_{\infty} \le 1.21 \cdot 10^3$ 



MATLAB<sup>®</sup> demo.

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MATLAB<sup>®</sup> demo.

# Basic Principle

#### Base enrichment

Static modes are defined by setting  $\dot{x} = 0$  and assuming unit loads, i.e.,  $u(t) \equiv e_j, j = 1, ..., m$ :

$$0 = Ax(t) + Be_j \implies x(t) \equiv -A^{-1}b_j.$$

Projection subspace  $\mathcal{V}$  is then augmented by  $A^{-1}[b_1, \dots, b_m] = A^{-1}B$ . Interpolation-projection framework  $\implies G(0) = \hat{G}(0)!$ 

If two sided projection is used, complimentary subspace can be augmented by  $A^{-T}C^T \Longrightarrow G'(0) = \hat{G}'(0)!$ 

Note: if  $m \neq q$ , add random vectors or delete some of the columns in  $A^{-T}C^{T}$ .

MOR by Projection

# Basic Principle

#### Guyan reduction (static condensation)

Partition states in masters  $x_1 \in \mathbb{R}^r$  and slaves  $x_2 \in \mathbb{R}^{n-r}$  (FEM terminology) Assume stationarity, i.e.,  $\dot{x} = 0$  and solve for  $x_2$  in

$$0 = \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} B_1 \\ B_2 \end{bmatrix} u$$
  
$$\Rightarrow \quad x_2 = -A_{22}^{-1}A_{21}x_1 - A_{22}^{-1}B_2u.$$

Inserting this into the first part of the dynamic system

$$\dot{x}_1 = A_{11}x_1 + A_{12}x_2 + B_1u, \quad y = C_1x_1 + C_2x_2$$

then yields the reduced-order model

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$$\dot{x}_1 = (A_{11} - A_{12}A_{22}^{-1}A_{21})x_1 + (B_1 - A_{12}A_{22}^{-1}B_2)u y = (C_1 - C_2A_{22}^{-1}A_{21})x_1 - C_2A_{22}^{-1}B_2u.$$

#### Modal Truncation Dominant Pole Algorithm

## Pole-Residue Form of Transfer Function

Consider partial fraction expansion of transfer function with D = 0:

$$G(s) = \sum_{k=1}^{n} \frac{R_k}{s - \lambda_k}$$

with the residues  $R_k := (Cx_k)(y_k^H B) \in \mathbb{C}^{q \times m}$ .

MOR by Projection

#### Modal Truncation Dominant Pole Algorithm

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**Note:** this follows using the spectral decomposition  $A = XDX^{-1}$ , with  $X = [x_1, ..., x_n]$  the right and  $X^{-1} =: Y = [y_1, ..., y_n]^H$  the left eigenvector matrices:

$$G(s) = C(sI - XDX^{-1})^{-1}B = CX(sI - \operatorname{diag} \{\lambda_1, \dots, \lambda_n\})^{-1}YB$$

$$= [Cx_1, \dots, Cx_n] \begin{bmatrix} \frac{1}{s-\lambda_1} & & \\ & \ddots & \\ & & \frac{1}{s-\lambda_n} \end{bmatrix} \begin{bmatrix} y_1^HB \\ \vdots \\ y_n^HB \end{bmatrix}$$

$$= \sum_{k=1}^n \frac{(Cx_k)(y_k^HB)}{s-\lambda_k}.$$

MOR by Projection

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**Note:**  $R_k = (Cx_k)(y_k^H B)$  are the residues of G in the sense of the residue theorem of complex analysis:

$$\operatorname{res} (G, \lambda_{\ell}) = \lim_{s \to \lambda_{\ell}} (s - \lambda_{\ell}) G(s) = \sum_{k=1}^{n} \lim_{\substack{s \to \lambda_{\ell} \\ f = \lambda_{\ell}}} \frac{s - \lambda_{\ell}}{s - \lambda_{k}} \quad R_{k} = R_{\ell}.$$
$$= \begin{cases} 0 \text{ for } k \neq \ell \\ 1 \text{ for } k = \ell \end{cases}$$

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As projection basis use spaces spanned by right/left eigenvectors corresponding to dominant poles, i.e..  $(\lambda_i, x_i, y_i)$  with largest

 $||R_k||/|\operatorname{re}(\lambda_k)|.$ 

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#### Remark

The dominant modes have most important influence on the input-output behavior of the system and are responsible for the "peaks"' in the frequency response.

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Random SISO Example ( $B, C^T \in \mathbb{R}^n$ )



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Modal Truncation



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